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To cite this version:

HAL Id: jpa-00255077
https://hal.archives-ouvertes.fr/jpa-00255077
Submitted on 1 Jan 1997
High Frequency Mn-Zn Power Ferrites

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Abstract. Relationships between chemical composition, microstructure and core losses of Mn-Zn power ferrites are discussed. We report in this paper the influence on core losses of silica-calcia additions and Ti-substitutions for optimising a power ferrite from 25 kHz up to 2 MHz. Our results are well explained using core losses description including the domain walls resonance-relaxation loss contribution depicted in previous papers and confirm that it is essential to take this contribution into account when optimising low losses ferrites in specific conditions of temperature, frequency and induction, especially in a high frequency range.

1. INTRODUCTION

The new generation of power converters requires higher operating frequencies to improve efficiency and compactness. Consequently, new active and passive components are needed. Among the passive components, the largest are the inductive ones such as transformers or inductances. In the middle and high frequency regions, magnetic cores of transformers are made with ferrites because these oxides exhibit high resistivities in spite of their low saturation magnetization compared with metallic materials (iron or silicon-iron). In SMPS (Switching Mode Power Supplies) applications, the operating frequency starts from 25 kHz to go up to 2 MHz while the operating induction decreases from 200 mT to 50 mT with working temperatures around 100°C. Accordingly, power ferrites characteristics have to be determined considering the physical mechanisms which are involved in core losses. The aim of this paper is to discuss the relationships between the chemical composition, the microstructure and the parameters influencing core losses of a power ferrite such as the saturation magnetization $M_S$ and the total magnetic anisotropy.

2. GENERAL CONSIDERATIONS

Losses in magnetic materials have two main contributions: one is due to hysteresis and originates from the non-reversibility of the magnetization process, the other is dynamic and should be studied at a more macroscopic scale. For time harmonic fields, losses in magnetic materials can be depicted by a difference in phase between the external applied magnetic field and the resulting magnetization. Whatever its physical origin, this time lag is closely related to magnetization mechanisms which occur in the material. These mechanisms are the domain walls displacements and the spin rotations. Their relative importance depends on basic magnetic parameters such as the magnetocrystalline and the magnetoelastic anisotropies and the magnetic structure of the Weiss domains. For soft ferrites, the main contributions to core losses are:

- **hysteresis** loss mainly due to the non-reversing domains walls displacements. This contribution exists in all magnetic materials. For dynamic applications, hysteresis loss could be estimated as the area of the quasi-static hysteresis loop times frequency. Hysteresis loss can be reduced by decreasing the coercitive field. This can be achieved by increasing the saturation magnetization $M_S$, decreasing the internal stress, the magnetostriction and the magnetocrystalline anisotropy. For a polycrystalline ferrite, the porosity and mostly the intra-granular porosity must be kept as low as possible [1], [2].

- **eddy current** loss due to electronic movements to cancel the applied magnetic field. This kind of loss is a part of the dynamic contribution: it is magnified in metallic magnetic materials where the electrical conductivity is high and is often negligible for most ferrites. For all magnetic materials, eddy current loss can be reduced by increasing the dynamic resistivity. Moreover, eddy current loss depends on the structure of the Weiss domains, so for polycrystalline ferrites such as Mn-Zn power ferrites, eddy current loss is microstructure dependent [3].

- **resonance-relaxation** loss due to either the reversible high frequency displacement of the domain walls or to the magnetization rotation inside a domain. It is another important part of the dynamic contribution which could dominate all other contributions if user does not take care. A practical way for choosing a ferrite is to consider the complex permeability.
spectrum for determining the frequency range in which losses occur (cf. figure 2). We have shown in previous works ([4],[5]) that a relevant parameter to account for the efficiency of a (power) ferrite is the $\mu_s f_r$ product: the higher this product, the lowest the resonance-relaxation loss (cf. figures 1a and 1b). This is always verified on condition that the ferrite quality is good, that is to say that the chemical composition and the grain size are homogeneous.

3. CHOOSING OF A POWER FERRITE: RELEVANT PARAMETERS

Considering the permeability spectrum on figure 2, it is clear that when approaching 1 MHz, the resonance-relaxation loss will contribute to core losses. Below this frequency, the main contribution to core losses will be hysteresis loss and eddy current loss if the material exhibits a low resistivity.

Although it is microstructure dependent, the $\mu_s f_r$ parameter may be considered as an intrinsic limit for a magnetic material. Experimental values of about 5 GHz have been observed for high frequency Mn-Zn power ferrites [4],[5]. It is mainly connected to the saturation magnetization and to the magnetic structure. Some authors have tried to determine theoretically this parameter for polycrystalline ferrites according to the magnetization mechanisms involved [6],[7],[8]. Their models lead to the following relations:

$$\mu_s f_r = \frac{12}{\beta D} M_s^2$$  in the case of domain wall motion [6],[7]

$$\mu_s f_r = \frac{\gamma}{3\pi} M_s$$  in the case of spin rotations [8]

where $\beta$ is the damping coefficient, $M_s$ is the saturation magnetization, $D$ is the average grain diameter and $\gamma$ is the gyromagnetic ratio. From these relations, it appears that a power ferrite, more generally any ferrite, should have a high saturation magnetization and a fine microstructure. When the grain size is small, the $\mu_s f_r$ product corresponding to the domain wall displacement is high and the spin rotations contribution is magnified [9]. Therefore the $\mu_s f_r$ product increases
whatever the magnetization mechanism involved. Figure 3 shows the permeability spectra for two Mn-Zn power ferrites: one has been prepared using the conventional ceramic route (class.), the other by the coprecipitation technology (coprec.). The main difference between these two ferrites is the grain size: the conventional ferrite has an average grain size between 6-7 μm whereas the coprecipitated one exhibits a fine microstructure with a lot of very small grains (1-2 μm). As a consequence, the coprecipitated ferrite has a lower static permeability and a higher resonance-relaxation frequency. Moreover, we observe a 30% reduction core losses at 1 MHz and 50 mT. As respectively the saturation magnetization and the dynamic resistivity are the same for both samples (resp. 0.6 T and 2 Ω.m), we conclude that resonance-relaxation loss is much lower for the coprecipitated sample. Under 3 MHz, the real part of the permeability is lower for the coprecipitated sample, after the contrary is observed. From 800 kHz, the coprecipitated sample exhibits lower losses compared with the conventional one.

The complex permeability of a conventional high frequency power ferrite has been measured under special biasing conditions (cf. figure 5): parallel biasing conditions where the static biasing field \( H \) is parallel to the alternative measurement field \( h \) (cf. fig. 4b) and perpendicular biasing conditions where the static biasing field \( H \) is perpendicular to the alternative measurement field \( h \) (cf. fig. 4a). The \((μ_μ^{-1})_{\mu} \) product remains practically constant when the applied static biasing field is parallel to the alternative measurement field whereas it increases when the static biasing field is perpendicular. This could be explained considering that perpendicular biasing magnify the spins rotation contribution whereas parallel biasing magnify the domain wall contribution as it has already been observed for other samples [9]. As a consequence, the same sample may have different dynamic behavior according to the biasing direction (cf. figure 5): \((μ_μ^{-1})_{\mu} = 5 \) GHz in the parallel configuration when \((μ_μ^{-1})_{\mu} = 10 \) GHz in the perpendicular configuration. Then, a Mn-Zn power ferrite presents better dynamic properties when the magnetization mechanisms involved are mainly spins rotation.

The total magnetic anisotropy of a ferrite which is the sum of the magnetocrystalline and the magnetoelastic anisotropies should be optimised according to the operating frequency range for preventing resonance-relaxation loss. Some authors have shown that Ti-substitution is a good way for both adjusting the magnetocrystalline anisotropy and increasing the dynamic resistivity of Mn-Zn [10]. However, this is not sufficient when operating at high frequency and some additives such as CaO to increase the resistivity and SiO2 to increase the density are necessary [11]. Table 1 summarizes the effects of microstructure, additions and substitutions on the different contributions to core losses. More details are given in the next paragraph.

<table>
<thead>
<tr>
<th></th>
<th>Hysteresis loss</th>
<th>Eddy current loss</th>
<th>Relaxation loss</th>
</tr>
</thead>
<tbody>
<tr>
<td>small grain size</td>
<td>high</td>
<td>low</td>
<td>low</td>
</tr>
<tr>
<td>(CaO-SiO2)</td>
<td>high</td>
<td>low</td>
<td>low</td>
</tr>
<tr>
<td>Ti-substitutions</td>
<td>low</td>
<td>low</td>
<td>high or low</td>
</tr>
</tbody>
</table>

Table 1: Relative effects of microstructure, additions and substitutions on the contributions to losses of a Mn-Zn power ferrite

4. CHEMICAL COMPOSITION, RAW MATERIALS AND CERAMIC PROCESS

The chemical composition and the microstructure of a power ferrite have to be chosen according to the operating frequency range and induction level. In all cases, the ferrous content Fe\(^{2+}\) is essential for obtaining a high saturation magnetization but it decreases the electric resistivity by magnifying the hopping mechanism between Fe\(^{2+}\) and Fe\(^{3+}\). The ways of increasing the electrical resistivity of Mn-Zn power ferrites are:
- Ti$^{4+}$ substitution because Ti$^{4+}$ acts as an electrostatic trap by pinning the electrons at Fe$^{2+}$ sites. Its presence induces Fe$^{2+}$ cations for maintaining the electronic balance and thus allows the minimization of the anisotropy constant $K_1$ versus temperature. Like this, the minimum of core losses versus temperature can be adjusted. As Ti$^{4+}$ must take place in the solid solution of the spinel phase, TiO$_2$ has to be introduced before the first ball-milling (see figure 5).

- CaO addition because Ca segregates at the grain boundary which leads to a highly resistive grain boundary [12]. Moreover, Ca acts as a grain growth inhibitor which increases the resistivity by maximising the grain boundary surface and overcomes the formation of a duplex microstructure consisting of abnormally large grains embedded in a fine grain matrix. We have shown in previous works ([5], [11]) that such a microstructure is responsible for relaxation loss due to the displacements of the very large domain walls at low frequency.

- SiO$_2$ addition because silica is acting as a flux: it improves the density and may increase the average grain size if firing conditions are not optimised [11]. The silica content must be carefully controlled because above 0.02 wt%, discontinuous grain growth takes place and core losses increase whereas the density is high [11]). Thus, for synthesizing a power ferrite with high performances using the conventional ceramique route, the impurities content must be known precisely. Moreover, the powder reactivity before firing has to be adjusted (specific area and oxidation state).

In the middle frequency range (32 kHz - 100 kHz), the grain size of power ferrites must be homogeneous and the optimum average grain size should be between 10 and 20 µm. If the grain size is lower, the magnetocrystalline anisotropy should be lowered to adjust the resonance-relaxation frequency so saturation magnetization will decrease. In the high frequency range (300 kHz - 2 MHz), the grain size of power ferrites must still be homogeneous but the optimum grain size should be between 5 and 10 µm. So firing conditions must be determined for each power ferrite according to the frequency range: for middle frequency, firing temperature should be high (1300 to 1350°C) and for high frequency, firing temperature should be lower (1200 to 1250°C) and firing time should be reduced. Moreover, during the firing, the oxygen partial pressure p$_{O_2}$ should be controlled to fix the right Fe$^{2+}$/Fe$^{3+}$ ratio, particularly during cooling [13]. In previous works [4], we observed that complex permeability and core losses minimum versus temperature were dependent on the oxidation degree and that it was related to the influence of Fe$^{3+}$/Fe$^{2+}$ content upon the magnetocrystalline anisotropy.

Using these considerations, a new high frequency power ferrite has recently been developed (reference F6). This ferrite could be used up to 2 MHz and presents the following characteristics:

<table>
<thead>
<tr>
<th>Static permeability</th>
<th>Density (g/cm$^3$)</th>
<th>Core losses at 1 MHz - 50mT</th>
<th>Core losses at 2 MHz - 40mT</th>
<th>Grain size (µm)</th>
<th>$\mu_e f_r$ product (GHz)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1000</td>
<td>4.7</td>
<td>470</td>
<td>1200</td>
<td>6 ± 0.5</td>
<td>5.6</td>
</tr>
</tbody>
</table>

5. CONCLUSIONS

The optimisation of middle and high frequency MnZn power ferrites can be achieved using controlled additions and substitutions. CaO additions are useful for lowering the ac-conductivity and allows the average grain size minimization. SiO$_2$ additions improve the ceramic density without damaging the electromagnetic properties up to a 0.02 wt%. TiO$_2$ substitutions allow the adjustment of the total magnetic anisotropy. Understanding these influences has enabled to develop a new high frequency material having low losses between 0.5 and 2 MHz.

References