Discrete and intersample analysis of systems with aperiodic sampling
Laurentiu Hetel, Alexandre Kruszewski, Wilfrid Perruquet, Jean-Pierre Richard

To cite this version:
Laurentiu Hetel, Alexandre Kruszewski, Wilfrid Perruquet, Jean-Pierre Richard. Discrete and intersample analysis of systems with aperiodic sampling. IEEE Transactions on Automatic Control, Institute of Electrical and Electronics Engineers, 2011, 56 (7), pp.1. <10.1109/TAC.2011.2122690>. <hal-00576366>
Discrete and intersample analysis of systems with aperiodic sampling

L. Hetel, A. Kruszewski, W. Perruquetti, J.P. Richard

Abstract

This article addresses the stability analysis of linear time invariant systems with aperiodic sampled-data control. Adopting a difference inclusion formalism, we show that necessary and sufficient stability conditions are given by the existence of discrete-time quasi-quadratic Lyapunov functions. A constructive method for computing such functions is provided from the approximation of the necessary and sufficient conditions. In practice, this leads to sufficient stability criteria under LMI form. The inter-sampling behavior is discussed there: based on differential inclusions, we provide continuous-time methods that use the advantages of the discrete-time approach. The results are illustrated by numerical examples that indicate the improvement with regard to the existing literature.

Index Terms

aperiodic sampled-data control, difference inclusions, stability, quasi-quadratic functions.

I. INTRODUCTION

The stability analysis of linear time invariant (LTI) systems with aperiodic sampling is a very challenging question. This problem is not easy, since, under variations of the sampling interval, the trajectory of a system may become unstable (see [23], page 69). The problem is relevant to networked / embedded control applications and has been addressed from both the discrete-time and continuous-time points of view.

In continuous-time, it has been approached using a time-delay system modeling [7], [6], a norm-bounded uncertainty modeling of the sample-and-hold operator [18], [9] or an impulsive...
system model [20]. The disadvantage of continuous-time methods is that, in general, the analysis does not take into account the particular variation of the sampling-induced delay which exhibits a “sawtooth shape”. The only method that considers this issue is the recent work [6]. Continuous-time methods may also suffer from conservatism due to the upper-bounding of the derivatives of Lyapunov-Krasovskii functionals or to the symmetry of ellipsoidal norms used for bounding the sample-and-hold operator.

In the discrete-time domain, using the exact integration over a sampling interval, the system with aperiodic sampling can be expressed as a linear difference inclusion (LDI): see [21], [5] for a switched LDI, [13] for a polytopic LDI, [1], [8] for a norm-bounded LDI and [22], [19], [14], [4] for generic stability properties of LDI. Sufficient linear matrix inequality (LMI) conditions for stability are given in the literature, based on the existence of quadratic Lyapunov functions [21], [5], [8], or of Lyapunov functions that depend on the sampling interval [13], [3]. Compared to continuous-time approaches, discrete-time methods profit by involving an integration procedure that implicitly takes into account the particular nature of the sampling-induced delay. The main drawback is that they cannot take into account the intersample system behavior. Besides, they become numerically inaccurate when the minimum sampling interval tends to zero.

This paper concerns both discrete-time (Section III) and continuous-time (Section IV) approaches for the stability analysis of LTI systems under aperiodic sampling. The LDI model suggests that improvement in the stability analysis can be made by using quasi-quadratic Lyapunov functions [19], [14], [24] which are both necessary and sufficient for the stability of LDI. However, the computation of such functions is still an open problem. To the authors knowledge, there is no LMI criteria in the literature for deriving such functions. This paper provides such a constructive numerical procedure for deriving quasi-quadratic Lyapunov functions. The approach is based on a necessary and sufficient condition that underlines a converse Lyapunov theorem for the stability of LDI defined on compact sets (as the LDI obtained from systems with aperiodic sampling). In practice, some restrictions to sufficient conditions lead to a LMI characterization of the stability domain that can be favorable compared to the existing literature. It is possible to tune the amount of conservatism of these LMI according to the desired numerical complexity. In a second time, we analyze systems with aperiodic sampling from the continuous-time point of view. The goal is to propose a new approach that uses the advantages of the discrete-time methods, via the classical integration operator. The approach can be used for analyzing the
intersample system behavior.

This paper is organized as follows: Section II provides a mathematical formulation of the problem under study; Section III deals with the discrete-time analysis, while Section IV presents continuous-time results. Numerical examples are presented in Section V.

Notations : For a square symmetric matrix, $M \succ 0$ ($M \prec 0$) indicates that $M$ is positive (negative) definite. $\| \cdot \|_M$ denotes the ellipsoidal norm associated to a matrix $M = M^T \succ 0$, $\| x \|_M = \sqrt{x^T M x}$. By $\lambda_{\text{max}}(M)$ ($\lambda_{\text{min}}(M)$) we denote the maximum (minimum) eigenvalue of a square symmetric matrix $M$. For a given set $\mathcal{F}$, the symbol $\text{co}\mathcal{F}$ denotes the convex hull of the set. For $p \in \mathbb{N}$, $\mathcal{I}_p$ denotes the set $\{1, 2, \ldots, p\} \subset \mathbb{N}$. For $k \in \mathbb{N}$ and a compact set $\Theta$ we denote by $S_k(\Theta) = \{ \sigma : \sigma = \{ \theta_i \}_{i=0}^{k-1}, \theta_i \in \Theta, \forall i = 0, \ldots, k-1 \}$ the set of all $k$–length sequences with values in $\Theta$. By $\nabla_y V(x) := \lim_{\epsilon \to 0^+} \frac{V(x + \epsilon y) - V(x)}{\epsilon}$ we denote the directional derivative of a function $V(x)$ along the direction $y$.

II. MATHEMATICAL FORMULATION

Consider two positive integers $n$, $m$ and the matrices $A_c \in \mathbb{R}^{n \times n}$, $B_c \in \mathbb{R}^{n \times m}$. We are interested in the class of LTI systems $\dot{x}(t) = A_c x(t) + B_c u(t)$, $\forall t \in \mathbb{R}^+$, $x(t) = x_0 \in \mathbb{R}^n$, $\forall t \leq 0$. Here $x : \mathbb{R} \to \mathbb{R}^n$ represents the system state and $u : \mathbb{R} \to \mathbb{R}^m$ represents the control. We consider that the control law is a piecewise constant state feedback, i.e. $u(t) = K x(t_k)$, $\forall t \in [t_k, t_{k+1})$, where $\{t_k\}_{k \in \mathbb{N}}$ represents an unbounded monotonously increasing sequence of sampling instants with elements in $\mathbb{R}^+$, i.e.

$$0 = t_0 < t_1 < \ldots < t_k < \ldots; \quad t_k \in \mathbb{R}^+, \forall k \in \mathbb{N}; \quad \lim_{k \to \infty} t_k = \infty. \quad (1)$$

We denote by $\theta_k := t_{k+1} - t_k$ the sampling interval and we consider that for all $k$, $\theta_k$ belongs to a compact set $\mathcal{T} \subset \mathbb{R}^+$. The closed-loop system has the form

$$\dot{x}(t) = A_c x(t) + B_c K x(t_k), \quad \forall t \in [t_k, t_{k+1}), \quad x(t) = x_0 \in \mathbb{R}^n, \forall t \leq 0. \quad (2)$$

We denote by $\theta_{\text{min}}$ and $\theta_{\text{max}}$ the minimum and maximum of $\mathcal{T}$, respectively, $0 < \theta_{\text{min}} < \theta_{\text{max}} < \infty$. The problem under study is formulated as follows:

Problem : Is system (2) stable for all the possible sampling sequences satisfying $\theta_k \in \mathcal{T}$ and assumption (1) ?
The study in this paper will be grounded on the use of the integration operator \( \Lambda : \mathbb{R}^+ \to \mathbb{R}^{n \times n} \),
\[
\Lambda(\theta) := I + \int_0^\theta e^{sA_c} ds (A_c + B_cK).
\] (3)
The operator \( \Lambda(\cdot) \) is continuous w.r.t. \( \theta \), and \( \theta \) belongs to the compact set \( T \). Then the set \( \{ \Lambda(\theta) : \theta \in T \} \) is a compact subset of \( \mathbb{R}^{n \times n} \). We consider that there exists a polytopic set
\[
\mathcal{Z} = \text{co} \{ Z_1, Z_2, \ldots, Z_p \} \subset \mathbb{R}^{n \times n}
\] (4)
with a finite number \( p \) of vertices, so that \( \Lambda(\theta) \in \mathcal{Z} \), \( \forall \theta \in T \). Similarly, we consider the set
\[
\mathcal{W} = \text{co} \{ W_1, W_2, \ldots, W_p \} \subset \mathbb{R}^{n \times n}
\] (5)
with a finite number \( p \) of vertices, so that \( \Lambda(\theta) \in \mathcal{W} \), \( \forall \theta \in [0, \theta_{\text{max}}] \). Methods for the construction of such polytopic sets exist in the literature [12], [13], [11] and will not be discussed in this paper. In what follows, we use the integration operator and the convex sets (4), (5) for both a discrete-time and continuous-time analysis of systems with aperiodic sampling.

III. DISCRETE-TIME ANALYSIS

A. Generalities

In this section we recall some basic stability properties and the construction of a LDI model for systems with aperiodic sampling. For all \( t \in [t_k, t_{k+1}] \) the solutions of system (2) satisfy:
\[
x(t) = e^{(t-t_k)A_c}x(t_k) + \int_0^{(t-t_k)} e^{sA_c} ds B_cKx(t_k)
\] (6)
\[
= \left( I + \int_0^{(t-t_k)} e^{sA_c} ds A_c + \int_0^{(t-t_k)} e^{sA_c} ds B_cK \right) x(t_k) = \Lambda(t-t_k)x(t_k)
\] (7)
where \( \Lambda(\cdot) \) is given in (3). At the sampling instants \( t_k \), system (2) is described by the LDI:
\[
x^+ \in \mathcal{H}(x), \quad \mathcal{H}(x) = \{ y : y = \Lambda(\theta)x, \theta \in T \},
\] (8)
where \( x^+ \) represents the state value at \( t_{k+1} \), i.e. \( x(t_{k+1}) \in \mathcal{H}(x(t_k)) \). Consider the initial condition \( x_0 \in \mathbb{R}^n \) and a \( k \) – length sequence \( \sigma = \{ \theta_i \}_{i=0}^{k-1} \in S_k(T) \). For any \( r \in \mathbb{N}, \ r \leq k \), we denote by \( \phi_\sigma \) the flow \( (r, x_0) \mapsto \phi_\sigma(r, x_0) \) defined by \( \phi_\sigma(r, x_0) = \Phi_\sigma(r)x_0 \) where \( \Phi_\sigma(r) \) represents the \( r \) – step transition matrix of (8) associated to \( \sigma \):
\[
\Phi_\sigma(r) = \begin{cases} 
\Lambda (\theta_{r-1}) \ldots \Lambda (\theta_1) \Lambda (\theta_0), & r > 0 \\
I, & r = 0.
\end{cases}
\] (9)
The solution of (8) associated to the initial condition $x_0$ and to an infinite–length sequence $\sigma \in S_\infty(T)$ represents the sequence of points $\{\phi_\sigma(k, x_0)\}_{k=0}^\infty$.

**Definition 1:** The equilibrium point $x = 0$ of (8) is said to be globally uniformly exponentially stable if there are constants $c > 0$, $0 < \lambda_d < 1$ s.t.:

$$\|\phi_\sigma(k, x_0)\| \leq c\lambda_d^k\|x_0\|, \forall k \geq 0$$

holds for all initial conditions $x_0 \in \mathbb{R}^n$, all $k \in \mathbb{N}$ and all sequences $\sigma \in S_\infty(T)$.

It has been shown in [9] that the stability of the LDI (8) is equivalent to the stability of the continuous-time system (2). The simplest stability criterion is the existence of a quadratic Lyapunov function (a sufficient condition). Note that for LDI, this criterion is not necessary, and may be a conservative stability test [22], [14], [19]. Here we show how to refine the analysis of systems with aperiodic sampling by using Lyapunov functions that are more suitable for LDI.

**B. Stability analysis based on discrete-time quasi-quadratic Lyapunov functions**

In what follows, we introduce a new discrete-time stability characterization for systems with aperiodic sampling. Before presenting the approach, we give a technical result for the LDI (8). Several necessary and sufficient conditions exist for the stability of polytopic LDI [17], [15], [14]. However, in the case of sampled-data systems (2), the obtained LDI (8) is not described by a polytopic set of vector fields. It can just be said that the set of matrices described by $\{\Lambda(\theta) : \theta \in T\}$ is a compact subset of $\mathbb{R}^{n \times n}$. As follows, we provide necessary and sufficient stability conditions that underline a converse Lyapunov theorem for LDI defined on compact sets (such as the ones obtained in the case of sampled-data systems).

**Theorem 1:** Consider the continuous-time system (2) and the equivalent LDI model (8) at the sampling instants. The following statements are equivalent:

1) The equilibrium point $x = 0$ of (8) is globally uniformly exponentially stable.

2) There exist a matrix $P = P^T \succ 0$ and a positive integer $N$ s.t. the transition matrix $\Phi_\sigma(N)$ defined in (9) satisfies the relation

$$P \succ \Phi_\sigma^T(N)P\Phi_\sigma(N), \forall \sigma \in S_N(T).$$

3) There exists a positive definite function $V : \mathbb{R}^n \rightarrow \mathbb{R}^+$ strictly convex, homogeneous (of the second order), $V(x) = x^T\mathcal{L}[x]x$, with $\mathcal{L} : \mathbb{R}^n \rightarrow \mathbb{R}^{n \times n}$, $\mathcal{L}[x] = \mathcal{L}_a^T = \mathcal{L}_{[ax]}$, $\forall x \neq 0, a \in \mathbb{R}$.
\( R, a \neq 0 \) s.t. the following relation is satisfied:

\[
V(x) - \max_{\theta \in T} V(\Lambda(\theta)x) > 0. \tag{12}
\]

Moreover, for \( P \) and \( N \) satisfying relation (11) in 2), the function \( V \) in 3) is given by

\[
V(x) = x^T L[x]x, \text{ with } L[x] = \sum_{i=0}^{N-1} \Phi_{\sigma^*(x)}^T(i) P \Phi_{\sigma^*(x)}(i)
\]

where

\[
\sigma^*(x) = \arg \max_{\sigma \in S_{N-1}(T)} x^T \left( \sum_{i=0}^{N-1} \Phi_{\sigma}(i) P \Phi_{\sigma}(i) \right) x.
\]

The proof is given in the Appendix. The theorem is inspired from [17], [15]. Note that it however goes beyond [17], [15], where only the equivalence 1) \( \Leftrightarrow \) 2) has been established for polytopic LDI. Here we propose, for LDI defined on compact sets, a constructive way of obtaining Lyapunov functions from the inequalities (11). In the generic context of LDI, the theorem can be seen as an alternative to [16] where conditions for the existence of time-dependent Lyapunov functions were derived. The class of functions that we obtain here depends only on the system’s state and does not need information on the evolution of the time-varying parameters. The set of inequalities (11) represents a necessary and sufficient stability condition for systems with aperiodic sampling. However, in practice, the number of inequalities to be checked in (11) grows in an exponential manner according to \( N \). Therefore, finding a general solution for (11) represents an NP-hard\(^1\) problem [2], even for the simple case when the set \( T \) is finite. Still, one can reduce it to a simpler problem using a finite \( N \) and a convexification of the set of vector fields in (8), as follows:

**Theorem 2:** Consider system (2), the equivalent LDI (8), the set \( Z \) in (4) and the set

\[
\mathcal{Y}(Z) = \{ Y : Y = \prod_{i=0}^{N-1} Z_{\mu_i}, Z_{\mu_i} \in \mathcal{Z}, \mu_i \in \mathcal{I}_p \}.
\]

If there exist a positive integer \( N \) and a matrix \( P = P^T \succ 0 \) that satisfy

\[
P \succ Y^T PY, \forall Y \in \mathcal{Y}(Z), \text{ then}
\]

1) the equilibrium point \( x = 0 \) of (8) is globally asymptotically stable;
2) there exists a quasi-quadratic Lyapunov function with the form

\[
V(x) = \max_{i \in \mathcal{I}_M} x^T L_i x
\]

\(^1\)There is no numerical algorithm that is able to solve the problem in a polynomial time.
which is strictly decreasing along the solutions of (8), where \( L_i, \ i \in \mathcal{I}_M, \ M = p^{N-1} \), are obtained using an enumeration of the elements in the set

\[
\Omega = \left\{ Q^Z_\sigma(N) : Q^Z_\sigma(N) = \sum_{j=1}^{N-1} \left( \Pi_{i=1}^j Z_{\mu_i} \right)^T P \left( \Pi_{i=1}^j Z_{\mu_i} \right) + P, \ \sigma = \{ \mu_i \}_{i=1}^{N-1} \in S_{N-1}(\mathcal{I}_p) \right\}.
\]

The proof is given in the Appendix. The test involves a finite number of LMI that are sufficient for stability. In Section V, we provide numerical examples where the proposed test is less conservative than the existing ones based on quadratic Lyapunov functions (such as in [8]), or on poly-quadratic stability [13]. Note that the case of quadratic Lyapunov functions is included in the conditions provided here ((11) or (14)) since it corresponds to \( N = 1 \). The accuracy of the stability characterization from conditions (14) mainly depends on two factors: the length \( N \) of the horizon of analysis, and the accuracy of the polytopic embedding \( Z \) described in (4) (for more details on such convex embedding see [12], [13], [11]). The amount of conservatism introduced in the approach can be tuned according to these parameters.

**IV. CONTINUOUS-TIME ANALYSIS**

We now provide a characterization of the intersample behavior of (2). Note that (2) is a **differential equation with discontinuous right hand side**, since \( \forall (t - t_k) \in T: \)

\[
\frac{dx(t)}{dt} \in \Xi(t, t_k), \ \Xi(t, t_k) = \{ (A_c + B_c K) x(t), \ A_c x(t) + B_c K x(t_k) \}.
\] (16)

The first element \((A_c + B_c K) x(t)\) corresponds to an actuation instant, whereas \(A_c x(t) + B_c K x(t_k)\) corresponds to the intersample dynamics. The model (16) allows for studying the existence of a continuous-time Lyapunov function

\[
V(x) = x^T L[x] x = \max_{i \in \mathcal{I}_M} x^T L_i x,
\] (17)

with \( L_i, \ i \in \mathcal{I}_M \), a set of symmetric positive definite matrices, as stated in Theorem 3:

**Theorem 3**: Consider system (2), the equivalent model (16) and the polytopic set \( \mathcal{W} \) given in (5). If there exist a scalar \( \lambda > 0 \), a set of scalars \( \beta_{ij} \geq 0, \ i, j = 1, \ldots, M \), and matrices \( G_1, G_2 \in \mathbb{R}^{n \times n} \) s.t.:

\[
\begin{pmatrix}
A^T L_i + L_i A_c + \lambda L_i - \sum_{i \neq j} \beta_{ij} (L_j - L_i) + G_1 + G_1^T & L_i B_c K - G_1 W_1 + G_2^T \\
K^T B_c^T L_i - W_1^T G_1^T + G_2 & -G_2 W_1 - W_1^T G_2^T
\end{pmatrix} < 0,
\] (18)
for all \(i, j \in I_M\), and for all \(l \in I_p\), then the maximal directional derivative of the function (17) following the vector fields of the closed-loop system \(U(t, t_k) = \max_{y(t) \in \Xi(t, t_k)} \nabla_y V(x(t))\) satisfies the relation

\[
U(t, t_k) \leq -\lambda V(x(t)), \quad \forall t \in [t_k, t_{k+1}], \quad \forall x(t) \neq 0,
\]

and the origin \(x = 0\) of the closed-loop system (2) is exponentially stable. ■

The proof is given in the Appendix. The goal of the previous theorem is two-fold. First, it can be used for estimating the decay rate of a given quasi-quadratic Lyapunov function (17) over the sampling intervals. If the Lyapunov function is computed from the discrete-time method (Theorem 2), then the \(L_i\) matrices are given and the computation of the decay rate leads to a set of LMI. Second, the theorem can be applied for directly computing a Lyapunov function without using the discrete-time analysis. The proposed test leads to Bilinear Matrix Inequalities (BMI). However, they can be reduced to LMI by considering quadratic Lyapunov functions, which gives a bit more conservative stability conditions. The previous theorem provides a description of the state behavior at all instants of time (including the intersample behavior): if the set of matrix inequalities (18) are satisfied, then (19) holds, which implies that \(V(x(t)) < e^{-\lambda t} V(x_0), \quad \forall t > 0\). This is the same as \(\|x(t)\|^2 < c e^{-\lambda t} \|x_0\|^2\) with \(c = \frac{\max_{i \in I_M} \lambda_{\max}(L_i)}{\min_{i \in I_M} \lambda_{\min}(L_i)}\).

Remark 1: Conditions (18) are not feasible in the dead-beat control case, where for some \(\theta \in [0, \theta_{\max}]\), \(\Lambda(\theta)\) has eigenvalues at zero. Excluding this case, the existence of functions of the form \(V(x) = x^T L_\theta x\) is not only sufficient, but also necessary for the stability of (2). This can be easily shown since in this case the matrix \(\Lambda(\theta)\) is invertible for all \(\theta \in [0, \theta_{\max}]\). Then \(x(t_k) = \Lambda^{-1}(t - t_k)x(t)\) and (2) can be expressed as the linear differential inclusion:

\[
\frac{dx}{dt} \in H_c(x), \quad H_c(x) = \{ y : y = (A_c + B_c K \Lambda^{-1}(\theta)) x, \ \theta \in [0, \theta_{\max}] \}.
\]

For such differential inclusion, it is known that the existence of a quasi-quadratic Lyapunov function is necessary for stability [19], [14]. However, due to the matrix inversion that appears in (20), no efficient numerical tool exists for applying the results in [14] to (20). ■

In the following theorem, we restrict to asymptotic stability and propose an analysis method that includes the dead-beat case using the Razumikhin method (see for example [10]).

**Theorem 4:** Consider system (2) and the polytopic set \(\mathcal{W}\) given in (5). If there exist a matrix...
\[ P = P^T > 0, \text{ a scalar } \epsilon > 0, \text{ and matrices } G_1, G_2 \in \mathbb{R}^{n \times n} \text{ s.t.:} \]

\[
\begin{pmatrix}
A_c^T P + P A_c + G_1 + G_1^T + \epsilon P & P B_c K - G_1 W_l + G_2^T \\
K^T B_c^T P - W_l^T G_1^T + G_2 & -G_2 W_l - W_l^T G_2^T - \epsilon P
\end{pmatrix} \prec 0,
\]

\(\forall l \in I_p,\) then the origin \(x = 0\) of the closed loop system (2) is asymptotically stable. \(\blacksquare\)

The proof is given in the Appendix. Note that the set of conditions (21) represents an optimization problem which can be solved using a line search algorithm and LMI solvers. The theorem ensures that, within the sampling interval, the Lyapunov-Razumikhin function \(V(x) = x^T P x\) is always less than the value at the sampling instants, although it is not monotonously decreasing. It can be shown, using numerical examples, that this new approach provides a less conservative stability condition in comparison with the existing continuous-time approaches (see Example 2, in Section V). In fact, this stability test is comparable to the one provided in discrete-time by a common quadratic Lyapunov function. The advantage, w.r.t. the discrete-time approach, is the fact that the intersample behavior is explicitly taken into account and that a sampling interval tending to zero may be considered as well. However, in comparison with Theorem 3, it cannot be used for computing a decay rate over the sampling interval, it only ensures asymptotic stability, not exponential stability.

V. Numerical examples

**Example 1.** Consider a LTI system (2) described by:

\[ A_c = \begin{pmatrix} -0.5 & 0 \\ 0 & 3.5 \end{pmatrix}, \quad B_c = \begin{pmatrix} 1 \\ 1 \end{pmatrix} \quad \text{and} \quad K = \begin{pmatrix} 1.02 & -5.62 \end{pmatrix}. \]

\(\Lambda(\theta)\) in (3) is Schur for any sampling interval \(\theta \in [0, 0.46].\) However, switching among different values of \(\theta\) in this interval may lead to an unstable behaviour: one can notice that although both \(\Lambda(0.25)\) and \(\Lambda(0.45)\) are Schur, the transition matrix \(\Phi = \Lambda(0.25)\Lambda(0.25)\Lambda(0.45)\) has the eigenvalues outside the unit circle. This implies that when the sampling period is varying in a periodic pattern \(0.25 \rightarrow 0.25 \rightarrow 0.45 \rightarrow 0.25 \rightarrow 0.25 \rightarrow 0.45 \ldots,\) the closed-loop system is unstable. A similar unstable behavior can be observed for \(\theta \in \{0.1, 0.43\}\) since the transition matrix \(\Phi = (\Lambda(0.1))^6 \Lambda(0.43)\) is not Schur. We consider that the sampling interval arbitrary switches among the values \(\{0.1, \theta_{max}\}\) and we use Theorem 2 to compute the maximum \(\theta_{max} \in [0.1, 0.46]\) ensuring stability. Using the set of LMI (14), it is possible to find a quasi-quadratic
Lyapunov function of the form (15) for $N = 7$ up to $\theta_{max} = 0.41$ (which is very close to the value 0.43 for which an unstable sampling path exists). For $\theta_{max} = 0.41$, using the existing LMI solvers, it is impossible to find a common quadratic Lyapunov function [21], [8], [1] or a poly-quadratic one [13]. In fact, the maximum values of $\theta_{max}$ that can be obtained from quadratic and poly-quadratic Lyapunov functions are $\theta_{max} = 0.36$ and $\theta_{max} = 0.39$, respectively.

Example 2. Consider a continuous-time system described by the following matrices:

$$A_c = \begin{pmatrix} 1 & 15 \\ -15 & 1 \end{pmatrix}, \quad B_c = \begin{pmatrix} 1 \\ 1 \end{pmatrix}, \quad K = (5.33 \ - 9.33).$$

In order to construct a polytopic set $\mathcal{W}$ for $\Lambda(\theta)$, we use the method proposed in [13] based on a Taylor series expansion. We use a uniform partition of the interval $[0, \theta_{max}]$ into 10 subintervals and apply locally the embedding method (4th order development). Using Theorem 3, a continuous-time quadratic Lyapunov function can be found up to $\theta_{max} = 0.09$ (see Figure 1). For this example the integration operator is singular for $\theta \approx 0.092$ (see Remark 1) which shows that the obtained $\theta_{max}$ is close to the theoretical bound for quasi-quadratic functions.

The methods in [18], [20], [9] and [6] show that the system is stable for $\theta_{max} = 0.014$, $\theta_{max} = 0.033$, $\theta_{max} = 0.07$ and $\theta_{max} = 0.12$, respectively. Theorem 4 proves the asymptotic stability for $\theta \in [0, 0.14]$ which is less conservative than the existing approaches. Note that using the discrete-time approach (Theorem 2), we are able to show the stability for $\theta \in [0.001, 0.15]$.
This means that Theorem 4 is almost as efficient as the discrete-time approach, with the additional advantage that it takes into account the intersample behavior and very small sampling intervals. Comparing now the number of LMI decision variables, [18] and [9] have 0.5(n^2 + n) + m^2 + m = 5 variables, [20] has 3.5n^2 + 1.5n = 17 while [6] has 8n^2 + n = 34. In Theorem 4 there are 0.5(n^2 + n) + 2n^2 = 11 variables involved in p + 1 = 51 LMI constraints.

VI. CONCLUSIONS

In this paper we addressed the robust stability of LTI systems with aperiodic sampled-data controller. The approach is based on LMI conditions for the existence of quasi-quadratic Lyapunov functions. It may also be of interest for a more general class of LDI. The intersample behavior is analyzed through a continuous-time analysis that benefits from the discrete-time approach. The results are illustrated by numerical examples that indicate improvement with regard to other recent approaches.

APPENDIX

Proof of Theorem 1: 1) ⇒ 2) Assume that (10) holds. Then there exists a symmetric positive definite matrix P and the associated ellipsoidal norm ||·||_P s.t. ||φ_σ(k, x)||_P^2 ≤ λmax(P)2^2k ||x||_P^2 holds for all x ∈ R^n, all k ∈ N and all σ ∈ S_∞(T). Since λ_d < 1, there exists a positive integer N s.t. ||φ_σ(N, x)||_P^2 < ||x||_P^2. This is the same as φ_T(N, x)Pφ_σ(N, x) − xTPx < 0, ∀x ∈ R^n, ∀σ ∈ S_∞(T). With φ_σ(N, x) = φ_σ(N)x, this is the same as R_σ(P, N) = Φ_T(N)PΦ_σ(N) − P < 0 ∀σ ∈ S_∞(T). Since R_σ(P, N) depends only on the first N terms of σ, then R_σ(P, N) < 0, ∀σ ∈ S_N(T), which implies (11).

2) ⇒ 3) Consider a sequence σ ∈ S_{N−1}(T). Define

Q_σ(N) = \sum_{i=0}^{N-1} Φ_T(i)PΦ_σ(i). \tag{22}

Q_σ(N) is continuous with respect to σ ∈ S_{N−1}(T). Since S_{N−1}(T) is a compact set, then sup_{σ∈S_{N−1}(T)} x^TQ_σ(N)x = max_{σ∈S_{N−1}(T)} x^TQ_σ(N)x. Thus V(x) = max_{σ∈S_{N−1}(T)} x^TQ_σ(N)x is homogeneous and convex. Furthermore, P = P^T ⪰ 0 implies Q_σ(N) = Q_T(σ)^T(N) ⪰ 0.

We are going to show that if the point 2) is satisfied, then the increment of V(x) over one step satisfies

ΔV(x) = \max_{σ∈S_{N−1}(T)} (x^+)^TQ_σ(N)x^+ − \max_{σ∈S_{N−1}(T)} x^TQ_σ(N)x < 0 \tag{23}
with \( x^+ = \Lambda(\theta)x \). Consider two \((N - 1)\) - length sequences \( \alpha = \{ \alpha_i \}_{i=0}^{N-2}, \beta = \{ \beta_i \}_{i=0}^{N-2} \in S_{N-1}(T) \) defined by:
\[
\alpha = \arg \max_{\sigma \in S_{N-1}(T)} (x^+)^T Q_\sigma(N)x^+, \quad \beta = \arg \max_{\sigma \in S_{N-1}(T)} x^T Q_\sigma(N)x.
\]

From (23) we obtain
\[
\Delta V(x) = x^T \left( \Lambda^T(\theta) Q_\alpha(N) \Lambda(\theta) - Q_\beta(N) \right) x. \tag{24}
\]

From the definition of \( Q_\alpha(N) \), given in (22), with \( \sigma = \alpha \), one can notice that
\[
Q_\alpha(N) = \sum_{i=0}^{N-1} \Phi^T_\alpha(i) P \Phi_\alpha(i) = \Phi^T_\alpha(N-1) P \Phi_\alpha(N-1) + Q_\alpha(N-1). \tag{25}
\]

We denote by \( \gamma \) the sequence \( \gamma = \{ \gamma_i \}_{i=0}^{N-2} \in S_{N-1}(T) \) with \( \gamma_0 = \theta, \gamma_i = \alpha_{i-1}, i = 1, \ldots, N-2 \). Moreover, we denote by \( \delta \) the sequence \( \delta = \{ \delta_i \}_{i=0}^{N-1} \in S_N(T) \), with \( \delta_0 = \theta, \delta_i = \alpha_{i-1}, i = 1, \ldots, N-1 \). One can notice that the term \( \Lambda^T(\theta) Q_\alpha(N) \Lambda(\theta) \) in (24) can be expressed as
\[
\Lambda^T(\theta) Q_\alpha(N) \Lambda(\theta) = \Phi^T_\delta(N) P \Phi_\delta(N) + Q_\gamma(N) - P. \tag{26}
\]

Therefore the one-step increment of \( V(x) \) is given by
\[
\Delta V(x) = x^T \left( \Phi^T_\delta(N) P \Phi_\delta(N) + Q_\gamma(N) - P - Q_\beta(N) \right) x. \tag{27}
\]

Since \( \beta = \arg \max_{\sigma \in S_{N-1}(T)} x^T Q_\sigma(N)x \), one can see that \( x^T Q_\beta(N)x \geq x^T Q_\gamma(N)x \) for any \( \gamma \in S_{N-1}(T) \). Therefore \( \Delta V(x) \leq x^T \left( \Phi^T_\delta(N) P \Phi_\delta(N) - P \right) x \), which means that if the relation (11) is satisfied for all \( \delta \in S_N(T) \) then condition (12) holds.

**Proof of Theorem 2 :**

1) Since \( V \) is piecewise quadratic and \( Q_\sigma(N) = Q^T_\sigma(N) > 0 \), for all \( x \in \mathbb{R}^n \) one can find two positive scalars \( a, b \) s.t. \( a\|x\|^2 < V(x) < b\|x\|^2 \) with \( a = \inf_{\sigma \in S_N(T)} \lambda_{\min} (Q_\sigma(N)) \) and \( b = \sup_{\sigma \in S_N(T)} \lambda_{\max} (Q_\sigma(N)) \). If \( V(x^+) - V(x) < 0 \) for all \( x \in \mathbb{R}^n \), then there exist a positive scalar \( \epsilon < 1 \) s.t. \( V(x^+) < \epsilon V(x) \), which is the same as \( V(\phi_\sigma(k, x)) < \epsilon^k V(x) \), \( \forall x \in \mathbb{R}^n, \forall k > 0, \forall \sigma \in S_{\infty}(T) \). This leads to \( \|\phi_\sigma(k, x)\|^2 < \frac{b}{a} \epsilon^k \|x\|^2 \). \( \Box \)

2) This property can be shown by considering the LDI \( x^+ \in \mathcal{H}^e(x), \mathcal{H}^e(x) = \{ Z_i x, i \in I_p \} \), and showing that (15) is a Lyapunov function for this LDI, using similar arguments to the proof of 2) \( \Rightarrow \) 3) in Theorem 1. Indeed, one can notice that (14) implies that \( V(Z_i x) < V(x), \forall i \in I_p. \)
Since $V(x)$ is convex, $\sum_{i=1}^{p} \alpha_i V(Z_i x) \geq V(\sum_{i=1}^{p} \alpha_i Z_i x)$, for all positive scalars $\alpha_i$, $i \in \mathcal{I}_p$ s.t. $\sum_{i=1}^{p} \alpha_i = 1$. We obtain that $V(\sum_{i=1}^{p} \alpha_i Z_i x) < V(x)$. Moreover, since $\Lambda(\theta) \in \mathcal{Z}$, this leads to $V(\Lambda(\theta)x) < V(x)$, i.e. the function (15) is strictly decreasing along the system solutions. □

**Proof of Theorem 3:** Note that $\forall t \in [t_k, t_{k+1}]$, the maximal derivative of the function (17) along the solutions of (2) satisfies the relation: $U(t, t_k) \leq \max_{t-t_k \in [0, \theta_{max}]} \nabla (A_c x(t) + B_c K x(t_k)) V(x(t))$. The value of the maximal directional derivative of $V(x)$ along a vector $y$ is given by:

$$\nabla_y V(x) = \max_{i \in I(x)} 2x^T L_i y, \quad I(x) := \left\{ i^*(x) : i^*(x) = \arg \max_{i \in \mathcal{I}_M} x^T L_i x \right\}. \quad (28)$$

Then

$$U(t, t_k) \leq \max_{t-t_k \in [0, \theta_{max}]} \max_{i \in I(x)} \max_{t \in [0, \theta_{max}]} 2x(t)^T L_i \{ A_c x(t) + B_c K x(t_k) \} \quad (29)$$

Therefore, condition (19) is satisfied if there exists a $\lambda > 0$ s.t.

$$\max_{t-t_k \in [0, \theta_{max}]} \max_{i \in I(x)} \begin{pmatrix} x(t) \\ x(t_k) \end{pmatrix}^T \begin{pmatrix} A_c^T L_i + L_i A_c + \lambda L_i & L_i B_c K \\ L_i B_c K & K^T B_c^T L_i \end{pmatrix} \begin{pmatrix} x(t) \\ x(t_k) \end{pmatrix} < 0. \quad (30)$$

Equation (7) implies that for $(t - t_k) \in [0, \theta_{max}]$, the following relation is satisfied:

$$\begin{pmatrix} I & -\Lambda(t-t_k) \end{pmatrix} \begin{pmatrix} x(t) \\ x(t_k) \end{pmatrix} = 0. \quad (31)$$

Using Finsler’s lemma for all $(t - t_k) \in [0, \theta_{max}]$ and $i \in \mathcal{I}_M$, one can notice that the set of relations

$$\begin{pmatrix} x(t) \\ x(t_k) \end{pmatrix}^T \begin{pmatrix} A_c^T L_i + L_i A_c + \lambda L_i & L_i B_c K \\ L_i B_c K & K^T B_c^T L_i \end{pmatrix} \begin{pmatrix} x(t) \\ x(t_k) \end{pmatrix} < 0, \quad t - t_k \in [0, \theta_{max}], \quad i \in \mathcal{I}_M,$$

under the constraint (31) is satisfied if there exist matrices $G_1, G_2 \in \mathbb{R}^{n \times n}$ s.t.

$$\Psi_i (x(t), x(t_k)) = \begin{pmatrix} x(t) \\ x(t_k) \end{pmatrix}^T \begin{pmatrix} A_c^T L_i + L_i A_c + \lambda L_i & L_i B_c K \\ L_i B_c K & K^T B_c^T L_i \end{pmatrix} \begin{pmatrix} x(t) \\ x(t_k) \end{pmatrix} + \begin{pmatrix} x(t) \\ x(t_k) \end{pmatrix} \left\{ \begin{pmatrix} G_1 \\ G_2 \end{pmatrix} \begin{pmatrix} I & -\Lambda(t-t_k) \end{pmatrix} + \begin{pmatrix} I \\ -\Lambda^T(t-t_k) \end{pmatrix} \begin{pmatrix} G_1^T \\ G_2^T \end{pmatrix} \right\} \begin{pmatrix} x(t) \\ x(t_k) \end{pmatrix} < 0 \quad (32)$$

for all $\begin{pmatrix} x^T(t) \\ x^T(t_k) \end{pmatrix} \neq 0$, all $(t - t_k) \in [0, \theta_{max}]$ and all $i \in \mathcal{I}_M$. Therefore, the equation (30) is satisfied if $\max_{i \in I(x)} \Psi_i (x(t), x(t_k)) < 0$, $\forall (t - t_k) \in [0, \theta_{max}]$. Using (28), this is the same
as \( \psi_i(x(t), x(t_k)) < 0, \forall i \in I_M, \forall (t-t_k) \in [0, \theta_{\max}], \) s.t. \( x(t)^T (L_i - L_j) x(t) > 0, \forall j \in I_M, j \neq i. \) Applying the S-procedure leads to the equation

\[
\begin{pmatrix}
A_c^T L_i + L_i A_c + \lambda L_i - \sum_{i \neq j} \beta_{ij} (L_j - L_i) + G_1 + G_1^T & L_i B c K - G_1 \Lambda(\theta) + G_2^T \\
K^T B_c^T L_i - \Lambda^T(\theta)G_1 + G_2 & -G_2 \Lambda(\theta) - \Lambda^T(\theta)G_2^T
\end{pmatrix} < 0.
\]

Since \( \Lambda(\theta) \in \mathcal{W}, \forall \theta \in [0, \theta_{\max}] \), the previous relation is satisfied when condition (18) holds.

Note that the relation (19) implies that \( V(x(t)) < e^{-\lambda t} V(x_0) \forall t > 0 \), which is the same as \( \| x(t) \|^2 < c e^{-\lambda t} \| x_0 \|^2 \) with \( c = \frac{\max_{i \in I_M} \lambda_{\text{max}}(L_i)}{\min_{i \in I_M} \lambda_{\text{min}}(L_i)} \). □

**Proof of Theorem 4:** Consider the quadratic function \( V(x) = x^T P x, P = P^T > 0 \). Define the propositions (A): “\( \dot{V}(x(t)) < 0 \)”, (B): “\( \max_{\theta \in [0, \theta_{\max}]} V(x(t) - \theta) < \alpha V(x(t)) \)” and (C): “\( V(x(t_k)) < \alpha V(x(t)) \)”. According to the Razumikhin’s stability theorem, the trivial solution \( x = 0 \) is asymptotically stable if there exists a scalar \( \alpha > 1 \) and a function \( V(x) \) s.t. (B) ⇒ (A).

Note that (B) ⇒ (C), so it is sufficient to show that there exist a scalar \( \alpha > 1 \) and a matrix \( P = P^T > 0 \) s.t. (C) ⇒ (A). The corresponding conditions can be expressed as

\[
\begin{pmatrix}
x(t) \\
x(t_k)
\end{pmatrix}^T \begin{pmatrix}
A_c^T P + P A_c & PB_c K \\
K^T B_c^T P & 0
\end{pmatrix} \begin{pmatrix}
x(t) \\
x(t_k)
\end{pmatrix} < 0 \quad \text{and} \quad \begin{pmatrix}
x(t) \\
x(t_k)
\end{pmatrix}^T \begin{pmatrix}
-\alpha P & 0 \\
0 & P
\end{pmatrix} \begin{pmatrix}
x(t) \\
x(t_k)
\end{pmatrix} < 0,
\]

respectively. Using the S-procedure, the stability condition is satisfied if there exists \( \epsilon > 0 \) s.t.

\[
\begin{pmatrix}
x(t) \\
x(t_k)
\end{pmatrix}^T \begin{pmatrix}
A_c^T P + P A_c + \epsilon \alpha P & PB_c K \\
K^T B_c^T P & -\epsilon P
\end{pmatrix} \begin{pmatrix}
x(t) \\
x(t_k)
\end{pmatrix} < 0.
\]

Similarly to the proof of Theorem 3, (31) and the Finsler’s lemma, leads to

\[
\begin{pmatrix}
A_c^T P + P A_c + \epsilon \alpha P + G_1 + G_1^T & PB_c K - G_1 \Lambda(\theta) + G_2^T \\
K^T B_c^T P - \Lambda^T(\theta)G_1 + G_2 & -G_2 \Lambda(\theta) - \Lambda^T(\theta)G_2^T - \epsilon P
\end{pmatrix} < 0. \tag{33}
\]

Since \( \Lambda(\theta) \in \mathcal{W}, \forall \theta \in [0, \theta_{\max}] \), if the set of conditions (21) is satisfied, then there exists a sufficiently small positive \( \delta \) s.t. (33) is satisfied with \( \alpha = 1 + \delta \). □

**References**


