Driven-Voltage Permeability Variation Measurements of Bilayered Magnetostrictive/Piezoelectric Materials & for Tunable Microwave Applications

Serge De Blasi, Patrick Queffelec, Sébastien Dubourg, Olivier Bodin, Marc Ledieu

To cite this version:
Serge De Blasi, Patrick Queffelec, Sébastien Dubourg, Olivier Bodin, Marc Ledieu. Driven-Voltage Permeability Variation Measurements of Bilayered Magnetostrictive/Piezoelectric Materials & for Tunable Microwave Applications. IEEE Transactions on Magnetics, Institute of Electrical and Electronics Engineers, 2007, 43 (6), pp.2651. <10.1109/TMAG.2007.893789>. <hal-00172013>

HAL Id: hal-00172013
https://hal.archives-ouvertes.fr/hal-00172013
Submitted on 13 Sep 2007
Driven-Voltage Permeability Variation Measurements of Bilayered Magnetostrictive/Piezoelectric Materials & for Tunable Microwave Applications

S. De Blasi¹, P. Quéffélec¹, S. Dubourg², O. Bodin², and M. Ledieu²

¹LEST-UMR CNRS 6165, Université de Bretagne Occidentale, Brest Cedex 393837-29238, France
²CEA, Le Ripault, Monts 37260, France

The complex permeability variation measurements at microwave frequencies of a magnetostrictive/piezoelectric bilayer under different values of a dc electric field are reported. A 25% variation of the initial permeability is achieved under a 1.5×10⁶ V/m static electric field.

Index Terms—Ferromagnetic materials, magneto-mechanical effects, microwave tunable devices, piezoelectric materials.

I. INTRODUCTION

There is a great number of wireless communication applications working in the decimeter wave range, and the question a user may legitimately ask is whether or not his mobile phone will be able to operate with the different standards or if it will be necessary to get different systems each working at a specific frequency to cover the whole range. One solution is a single device incorporating various components, each working at a fixed frequency, but the best solution is a device incorporating microwave components able to work at various frequencies. Tunability which consists of modifying the transmission characteristics (center operating frequency and/or pass-bandwidth) thanks to an external command (static current or voltage) was previously achieved by inserting active components [1].

But, because of the well-known limitations of this approach [low signal-to-noise ratio (SNR)], substitute technologies like MEMS [2] and dielectric substrates exhibiting a variable permittivity such as ferroelectrics [3] have been studied. In the same way, a sensitive tunability was performed by inserting magnetic/dielectric composites into a transmission line under a weak static magnetic field [4], [5]. However, the integration of the magnetic control device (coils) is not compatible with the miniaturization of microwave circuits. Ferromagnetic materials would be good candidates for tunable applications at microwave frequencies, provided a new kind of static command can be developed. Fortunately, it is well known that the magnetic properties of a material under stress can also be varied (reverse magnetostrictive effect). The present paper aims at measuring how the magnetic properties, such as the permeability spectra (the permeability magnitude and the gyromagnetic resonance frequency), of a ferromagnetic layer are modified at microwave frequencies, when the piezoelectric substrate on which it lies is strained under the action of a static electric field (reverse piezoelectric effect).

II. ELECTRIC FIELD SENSITIVITY OF A MAGNETIC MATERIAL

A. Inverse Magnetostrictive Effect

A uniaxial stress $\sigma_x$ applied to a magnetic layer induces an equivalent stress-induced anisotropy field $H_{\text{as}}$ [6]. It can be defined as the value of the static magnetic field necessary to rotate in its direction the moments of a magnetic sample under stress

$$H_{\text{as}} = \frac{2K_1 - 3\lambda_0 \sigma_x}{\mu_0 M_S}$$

where $M_S$ is the saturation magnetization, $\mu_0$ the permeability of vacuum, $\lambda_0$ the magnetostriction constant, and $K_1$ the magneto-crystalline anisotropy energy. Then, the initial magnetic susceptibility can be expressed as

$$\chi(0) = \frac{\partial M}{\partial H} = \frac{\mu_0 M_S^2}{2K_1 - 3\lambda_0 \sigma_x}.$$  \hspace{1cm} (2)

The equation of motion without damping of a magnetic moment under the effective dc field $H_{\text{as}}$ [7] gives rise to an expression of the high frequency susceptibility

$$\chi(f) = \frac{\chi(0)}{1 - (f/f_0)^2}$$  \hspace{1cm} (3)

with $f_0 = \gamma \sqrt{H_{\text{as}}(H_{\text{as}} + 4\pi M_S)}$, where $\gamma$ is the gyromagnetic ratio.

Even if (3) predicts a variation under stress of the rf spin permeability in the whole frequency range, we will only consider in the next section the static permeability, because its variations implies a shift of the real part of the permeability spectrum below the gyromagnetic absorption region, which is aimed for magnetic-based tunable applications.

B. Inverse Piezoelectric Effect

The inverse piezoelectric effect [8] corresponding to the strain $\epsilon_k$ induced by a piezoelectric material under a static applied electric field $E_m$ can be expressed along the $x$ axis as

$$\epsilon_x = d_{11} E_x + d_{21} E_y + d_{31} E_z$$  \hspace{1cm} (4)

where the $d_{ij}$ are the piezoelectric coefficients, when no external stress is applied to the material.
In the case of a static electric field applied in the $z$ direction

$$e_x = d_{31}E_z$$

and

$$\sigma_x = Y d_{31}E_z$$  \hspace{1cm} (5)

where $Y$ is the Young’s modulus in the $(x, y)$ plane and $d_{31}$ the deformation in the $x$ direction when an electric field is applied in the $z$ direction.

C. Coupled Inverse Magnetostrictive/Piezoelectric Effects

If a magnetic layer set onto a piezoelectric substrate is biased by a static electric field, its properties will vary because of the stresses induced by the substrate deformation due to the inverse piezoelectric effect. Indeed, if (5) is inserted into (1), it follows that an equivalent electrical anisotropy field $H_{aE}$ can be expressed as

$$H_{aE} = \frac{2K_1 - 3\lambda_s Y d_{31} E_z}{\mu_0 M_s}.$$  \hspace{1cm} (6)

In the same way, if (5) is inserted into (2), the magnetic susceptibility becomes

$$\chi_{\sigmaE} = \frac{\partial M}{\partial H} = \frac{\mu_0 M_s^2}{2K_1 - 3\lambda_s Y d_{31} E_z},$$

and the electric field sensitivity of the magnetic layer is

$$\frac{\partial \chi_{\sigmaE}}{\partial E} = \frac{3\lambda_s Y d_{31} \mu_0 M_s^2}{(2K_1 - 3\lambda_s Y d_{31} E_z)^2}.$$  \hspace{1cm} (7)

With the equation above, if the magnetic material has the same Young’s modulus and thickness as the piezoelectric substrate, the initial static susceptibility variation of a magnetic material, lying on a piezoelectric substrate under a static electric field, can be evaluated.

III. EXPERIMENTAL SETUP

A. Piezoelectric Substrate

Rectangular pieces (20 mm$^2$) of soft tetragonal Lead–Zirconate–Titane (PZT) were cut in a z-poled PPK-21 100-μm-thick piezoelectric substrate (STELCO GmbH). Silver pasted, screen printed, and burnt electrodes had been previously deposited on both sides of the substrate. This material offers great longitudinal $d_{33} = 500$ pC/N and transverse $d_{31} = -220$ pC/N piezoelectric constants, has a Curie temperature $\theta_C = 230^\circ$C, and has a Young’s modulus $Y' = 65,10^9$ N/m$^2$.

B. Ferromagnetic Layer

A soft ferromagnetic material will be preferred for microwave tunable applications because of its low coercive field, its high saturation magnetization to work over a wide frequency range (the gyromagnetic frequency $f_C \propto 4\pi M_s$), and its high static permeability allowing a significant shift under an external static command. Moreover, because the reverse magnetostrictive effect is to be used, the magnetic material must exhibit a high magnetostriction constant. Finally, because ferromagnetics are conductive materials, their thicknesses must also be thinner than the skin depth to avoid losses due to induced eddy currents. A 140-nm-thin Fe$_{66}$Co$_{18}$Si$_2$B$_{14}$ ferromagnetic amorphous layer with $\lambda_s = 35 \times 10^{-6}$ was first deposited by magnetron sputtering on a thick smooth glass substrate. We measured a $4\pi M_s = 17800$ G, an anisotropy field $H_K = 120$ Oe, and a resistivity $\rho = 140 \mu\Omega$ cm.

The permeability spectra show a natural (without external bias) gyromagnetic resonance at 4 GHz. A damping parameter $\alpha \approx 0.005$ was estimated. The real part of the permeability is close to 500, there are no contributions of domain wall movements (Fig. 1) and the imaginary part of the permeability (losses) is weak at $f = 2$ GHz. Thus, this magnetic material can be used in tunable telecommunication devices working near this frequency (wireless applications).

C. Ferromagnetic/Piezoelectric Composite

A 30-μm glass substrate was bonded as an under layer onto the top Ag electrode of the PZT substrate (Fig. 2) because the deposition of the ferromagnetic layer directly onto the rough piezoelectric substrate, using the conductive magnetic material as an electrode, did not give interesting results. Then, the 140-nm-thin Fe$_{66}$Co$_{18}$Si$_2$B$_{14}$ ferromagnetic amorphous layer was deposited by magnetron sputtering onto the glass under layer.

During the deposition, the samples were oriented to obtain a transverse easy axis of magnetization, and special care was taken not to exceed the Curie temperature of the PZT substrate to preserve its remanent polarization.
IV. MEASUREMENTS

A permeameter based on the measurement of the impedance of an inductive loop [9] containing the polarized ferromagnetic/piezoelectric composite to be characterized was used to demonstrate the variations of the real and imaginary parts of the permeability at microwave frequencies.

The frequency dependency of the losses plotted in Fig. 3(b), which corresponds to the magnetic material deposited as shown in Fig. 2, is less sharply defined than the one obtained with the glass substrate (Fig. 1). This difference can probably be explained by a nonuniformity of the local anisotropy field strength distribution, due to the higher roughness of the 30-μm-glass underlayer.

The real part of the permeability is increased of 25% at a frequency \( f = 2 \) GHz, and the resonance frequency is shifted from 4 down to 3.8 GHz under the action of a \( 1.5 \times 10^6 \) V/m static electric field. These results are in good agreement with Snoek’s law predicting an inverse relationship between the resonance frequency and the static permeability [10]. These variations are the consequence of a diminution of the net anisotropy in the transverse direction due to the strained piezoelectric substrate. According to (5), such a static electric field should correspond to an estimated net stress in the \( x \) direction of \( \sigma_x = 50 \times 10^6 \) N/m\(^2\) (\( V_{\text{ferro}} \approx 170 \times 10^6 \) N/m\(^2\) [11]). In fact, the deformation of the piezoelectric substrate yields a balanced biaxial stress in the \((x,y)\) plane inducing an anisotropy in the \((y,z)\) plane too, which decreases the stress induced anisotropy in the \((x,y)\) plane [12]. An electric field-induced uniaxial deformation of the PZT substrate would probably give greater permeability variations.

V. CONCLUSION

A 140-nm-thin ferromagnetic \( \text{Fe}_{44}\text{Co}_{18}\text{Si}_{2}B_{14} \) has been deposited on a soft 100-μm-thick PZT substrate, and a 25% variation of the magnetic layer permeability has been measured as a static electric field of \( 1.5 \times 10^5 \) V/m was polarizing the piezoelectric layer. By solving the problem of the insertion of a static magnetic command in electronic devices, this work restores ferromagnetic materials as very good candidates for the design of miniaturized microwave tunable devices for telecommunication applications. After a design optimization to get an uniaxial induced stress to perform the magnetic property variations, a future paper will show how optimized magnetostrictive/piezoelectric composites can be inserted in a transmission line to tune the characteristics of a microwave circuit.

ACKNOWLEDGMENT

This work was supported in part by the Brittany Region.

REFERENCES


Manuscript received October 31, 2006 (e-mail: serge.deblasi@univ-brest.fr).